

SBOA227A-February 2018-Revised January 2019

# Half-wave rectifier circuit

#### **Design Goals**

Input		Output		Supply	
V <sub>iMin</sub>	V <sub>iMax</sub>	V <sub>oMin</sub>	V <sub>oMax</sub>	V <sub>cc</sub>	V <sub>ee</sub>
$\pm 0.2 \text{mV}_{\text{pp}}$	±4V <sub>pp</sub>	0.1V <sub>p</sub>	2V <sub>p</sub>	2.5V	–2.5V

#### **Design Description**

The precision half-wave rectifier inverts and transfers only the negative-half input of a time varying input signal (preferably sinusoidal) to its output. By appropriately selecting the feedback resistor values, different gains can be achieved. Precision half-wave rectifiers are commonly used with other op amp circuits such as a peak-detector or bandwidth limited non-inverting amplifier to produce a DC output voltage. This configuration has been designed to work for sinusoidal input signals between  $0.2mV_{pp}$  and  $4V_{pp}$  at frequencies up to 50kHz.



#### **Design Notes**

- 1. Select an op amp with a high slew rate. When the input signal changes polarities, the amplifier output must slew two diode voltage drops.
- 2. Set output range based on linear output swing (see A<sub>ol</sub> specification).
- 3. Use fast switching diodes. High-frequency input signals will be distorted depending on the speed by which the diodes can transition from blocking to forward conducting mode. Schottky diodes might be a preferable choice, since these have faster transitions than pn-junction diodes at the expense of higher reverse leakage.
- 4. The resistor tolerance sets the circuit gain error.
- 5. Minimize noise errors by selecting low-value resistors.

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# Design Steps

1. Set the desired gain of the half-wave rectifier to select the feedback resistors.

 $V_o = \text{Gain} \times V_i$  $\text{Gain} = -\frac{R_f}{R_1} = -1$ 

 $R_{f}\!=R_{1}\!=2 \textbf{\times} R_{eq}$ 

- Where  $R_{\scriptscriptstyle eq}$  is the parallel combination of  $R_{\scriptscriptstyle 1}$  and  $R_{\scriptscriptstyle f}$
- 2. Select the resistors such that the resistor noise is negligible compared to the voltage broadband noise of the op amp.

$$\begin{split} \mathsf{E}_{\mathsf{nr}} &= \sqrt{4 \times \mathsf{k}_{\mathsf{b}} \times \mathsf{T} \times \mathsf{R}_{\mathsf{eq}}} \\ \mathsf{R}_{\mathsf{eq}} &\leq \frac{\mathsf{E}_{\mathsf{nbb}}^2}{4 \times \mathsf{k}_{\mathsf{b}} \times \mathsf{T} \times 3^2} = (\mathsf{E}\mathsf{nbb}) \\ &= 7 \cdot 5 \frac{\mathsf{nV}}{\sqrt{\mathsf{Hz}}} = \frac{(7.5 \times 10^{-9})^2}{4 \times 1.381 \times 10^{-23} \times 298 \times 3^2} = 380\Omega \\ \mathsf{R}_{\mathsf{f}} &= \mathsf{R}_1 \leq 760\Omega \to 750\Omega \text{ (Standard Value)} \end{split}$$

## **Design Simulations**

# **DC Simulation Results**





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 $2V_{\mbox{\tiny pp}}$  at 50kHz

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## **Design References**

See Analog Engineer's Circuit Cookbooks for TI's comprehensive circuit library.

See circuit SPICE simulation file SBOC509.

# Design Featured Op Amp

OPA322				
V <sub>ss</sub>	1.8V to 5.5V			
V <sub>inCM</sub>	Rail-to-rail			
V <sub>out</sub>	Rail-to-rail			
V <sub>os</sub>	500µV			
l <sub>q</sub>	1.6mA/Ch			
I <sub>b</sub>	0.2pA			
UGBW	20MHz			
SR	10V/µs			
#Channels	1, 2, 4			
www.ti.com/product/opa322				

## **Design Alternate Op Amp**

OPA2325				
V <sub>ss</sub>	2.2V to 5.5V			
V <sub>inCM</sub>	Rail-to-rail			
V <sub>out</sub>	Rail-to-rail			
V <sub>os</sub>	40µV			
l <sub>q</sub>	0.65mA/Ch			
I <sub>b</sub>	0.2pA			
UGBW	10MHz			
SR	5V/µs			
#Channels	2μ			
www.ti.com/product/opa2325				

## **Revision History**

Revision	Date	Change
А	January 2019	Downscale the title and changed title role to 'Amplifiers'. Added link to circuit cookbook landing page and link to Spice simulation file.